

MEWARP 2600

# TIMEWARP 2600

## OWNER'S MANUAL





## Owner's Manual



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# The ARP 2600, 1970 – 1981 and onward... 1

The ARP 2600 was the second product of ARP Instruments. It was released in 1970, and continued until the manufacturer ceased operations in 1981.

Its design combined *modularity* (for studio flexibility, and for instructional use) and *integration* (for realtime performance). Functionally, the ARP 2600 was completely modular: any signal output could be routed to any signal input, with a patch cord. Operationally, the ARP 2600 was integrated, using internally-wired default signal paths that made it possible to create a wide range of keyboard patches by simply opening up slide attenuators, as though sitting in front of a mixing console.

The ARP 2600 earned an early reputation for stability, and for flexibility exceeding that of its competitor the Minimoog. Used 2600's in good condition command premium prices on eBay today and businesses around the country can make a living reconditioning, rebuilding, and customizing 30-year-old units.

Among rock musicians, the ARP 2600 was used by Stevie Wonder, Pete Townsend, Joe Zawinul, Edgar Winter, Paul Bley, Roger Powell, Jean-Michel Jarre, Mike Oldfield, Herbie Hancock, and many, many others.

Its modular design, and the popularity of its Owner's Manual, made the ARP 2600 a widely used teaching instrument in many schools and music conservatories around the country and internationally.

We are proud to bring you this software emulation of the 2600. Have fun with it, learn from it, but above all, make noise with it.

## Foreword

Unfasten the seat belts of your mind. The TimewARP 2600 will be an astonishing, exhilarating, and enlightening experience.

Creating this manual has been an astonishing, exhilarating, and enlightening experience for *me*. How many are ever given the chance to revisit an earlier life, an earlier project, a project like the ARP 2600 Manual, decades later, and get it right? It's time travel. I'm grateful to Way Out Ware for providing me that opportunity.

When, at Alan R. Pearlman's invitation, I began work on the original 2600 manual in September of 1970, the 2600 itself barely existed. For the first two months, I was writing "blind" - without a machine in front of me. My first hands-on experience with a synthesizer had been only six months earlier (it was a Putney VCS3).

I finished the text in March of 1971, Margaret Friend created the graphics, and the Owner's Manual for the ARP 2600 began what turned out to be a surprisingly long career. In spite of the many defects that my inexperience contributed - the gaps in coverage, and outright errors - it became quite popular. To this day, it still gets an occasional respectable mention in the analog-synthesis community.

When Way Out Ware's Jim Heintz called, early in 2004, to tell me about the TimewARP 2600, a lot of time had passed. Regarding software synthesizers, I had grown weary and cynical. Analog-modeling software had been a decade-long disappointment; some products did interesting things but not the things that real analog modules do. Jim, however, had already encountered, and thought about, and solved, these problems. He owned a real 2600. He really aimed at getting it right and would not be satisfied with anything less. It was a pleasure, finally, to accept his invitation to do an Owner's Manual.

It's clear, now, that Way Out Ware has set a new standard for software-based audio synthesis. The behavior of the TimewARP 2600 software - both module-by-module and integrated into patches - is effectively indistinguishable from that of the analog hardware that it emulates.

Soaring and swooping through the free air of analog synthesis - a world of nothing but sliders and cords and continuously evolving patch configuration - was a capstone course at the Boston School of Electronic Music in the 1970's. That is the world that the TimewARP 2600, for a new generation of musicians in a new millennium (that means *you*), provides access to: it is the first - and I believe only - software synthesizer to support real-time performance by sliders and patchcords alone.

So here it is: your new Owner's Manual, for the new TimewARP 2600.

Unfasten the seat belts of your mind. How else can you hope to experience time travel? How else can you enjoy free flight?

*Jim Michmerhuizen*

*Jim Michmerhuizen is the author of the original ARP 2600 Manual and Founder and Director of the Boston School of Electronic Music.*

## How this Manual is Organized

This is not a textbook; it's a survival manual.

Chapter 2 is about installing and configuring the software so you can get up and running.

Chapter 3 is a brief introduction to the vocabulary and methods of classical analog synthesis, so that we can understand each other throughout the rest of the book.

Chapter 4 is a module-by-module reference, including the digital extensions made possible by the fact that the TimewARP 2600 is software – a piece of computer behavior – rather than a collection of electronic hardware.

Chapter 5 is not in this book, but is a collection of patches, with accompanying documentation, located in a separate document file called *Patchman.pdf* found on the distribution CD (or as a download from [www.wayoutware.com](http://www.wayoutware.com)). The patches and commentary are keyed to the numbered chapters, sections, and subsections of this manual. Some of the patches form a tutorial sequence, and some illustrate vocabulary lessons and concepts.

## How To Use This Manual

You'll probably need to do a quick run through of Chapter 2 as you install the TimewARP 2600 and learn some of the basic setup operations. After that, you can pretty much mix and match, according to your experience:

If you're new to audio synthesis, you'll want to walk through the tutorial patch sequence in *Patchman.pdf*, referring back to Chapter 3 for concepts and Chapter 4 for detailed module specifications while you explore, and learn to control, the vast range and patch repertoire of the TimewARP 2600 software synthesizer.

If you already have some experience with audio synthesis, you might go directly to Chapter 4 for the detailed module-by-module specifications of the TimewARP 2600. If you know and love the original ARP 2600 itself, take particular note of section 4.1, where we describe what you can do with the TimewARP 2600 that you could not do with its analog ancestor. These digital extensions include patch load/save, additional VCO sine-wave outputs, dual-channel signal input, automatic Y-connections at all signal outputs, sixteen keyboard micro-tuning options, MIDI Beat Clock synchronization, and MIDI controller mapping (with subranges) for all the panel sliders.



# System Requirements, Installation, Configuration, Setup and Usage 2

In this chapter you will find all of the platform-dependent information you need in order to install and operate your TimewARP 2600 software synthesizer.

The TimewARP 2600 provides an extended set of digitally-based features that no hardware-based analog synthesizer can offer. We describe the most important of these in section 2.2.

## **2.1 Digidesign Pro Tools**

This release of the TimewARP 2600 is an RTAS plug-in and runs only under Digidesign Pro Tools 6.1 and later versions, on either Apple OSX or Windows XP. On either of these platforms, the requirements for the TimewARP 2600 are essentially those of Pro Tools itself.

### **2.1.1 System Requirements**

These are the same as Pro Tools itself, including enough memory to satisfy the practical needs of the TimewARP 2600 software.

#### **2.1.1.1 Disk Space and RAM**

The executable file occupies 6.1MB on disc, and the RAM requirements when running are the same as the Pro Tools minimums.

#### **2.1.1.2 System Clock**

The TimewARP 2600 performs all signal generation and processing in real time. In complex polyphonic patches, this may put a considerable load on your CPU. You may have difficulty with a clock speed less than 800MHz; higher clock rates will increase the number of polyphonic voices you can use, and the complexity of the patches available.

### **2.1.2 Installation**

Installation on either platform is accomplished by very simple standardized installation sequences. You must, of course, have a current installation of Pro Tools on your target computer.

### 2.1.2.1 Apple OSX

Insert the distribution CD or download the .dmg installer file from [www.wayoutware.com](http://www.wayoutware.com). Double click the .dmg installer to initiate installation.

If you agreed to the license, a TimewARP 2600 installer icon appears. Double-click it to initiate the installation process.

When finished, start Pro Tools and proceed to 2.1.3.1.

### 2.1.2.2 Windows XP

Insert the distribution CD or download the installer file from [www.wayoutware.com](http://www.wayoutware.com). Run the .exe installer file.

If you agreed to the license, follow the installer instructions.

When finished, start Pro Tools and proceed to 2.1.3.1.

### 2.1.3 Setups

#### 2.1.3.1 Real-time Instrument for Performance and Recording

The TimewARP 2600 can transform your computer into a real-time instrument for live performance or recording.

To set this up, create an *audio* track in Pro Tools. Select either *mono* or *stereo* in the track-creation window; this will determine, in turn, the channel options offered by the TimewARP 2600 for this track.

In the mix window display for this track, at the track-inserts block up at the top, click on the first *insert* selector. Choose the TimewARP 2600 plug-in from the appropriate menu.

Now create another track, but in the track creation dialog, choose *MIDI track* instead of *audio track*, and route the track output to the TimewARP 2600.

As *input* for the new track, select your MIDI keyboard device, and enable the track for recording. Pro Tools will route all incoming MIDI events from your MIDI keyboard device to the TimewARP 2600 plug-in.

If your MIDI keyboard has any additional MIDI control devices (sliders, knobs, buttons, etc.), you can assign these to any TimewARP 2600 sliders, knobs and switches that you choose. For details, see section 2.2.

#### 2.1.3.2 Processing Audio Signals

The TimewARP 2600 can be used to process audio signals in a similar manner to a reverb or other effects plug-in.

To set this up, create an *audio* track or select a track already recorded in Pro Tools. (If this is a new track, select either *mono* or *stereo* in the track-creation window; this will determine, in turn, the channel options offered by the TimewARP 2600 for this track.)

In the mix window display for this track, at the track-inserts block up at the top, click on the first available *insert* selector. Choose the TimewARP 2600 plug-in from the appropriate menu.

If the current track is mono, the TimewARP 2600 for this track will be configured for one channel of input, and you can select between mono and stereo output. If the current track is stereo, the TimewARP 2600 insert will automatically be configured for stereo throughout. Confirm this, when the synthesizer panel comes up, by observing that the preamp module in the upper left corner has *two* channel outputs rather than just one.

If you are processing an existing track, press play on the transport control to hear the audio being processed by the TimewARP 2600 in real time. You need to select an appropriate TimewARP 2600 patch (such as those in the *Voice* or *Guitar Effects* categories in the *Factory* group) to successfully process audio this way.

If you are processing a live track, enable the track's record mode, and then audio from your input will be fed through the TimewARP 2600. You need to select an appropriate TimewARP 2600 patch (such as those in the *Voice* or *Guitar Effects* categories in the *Factory* group) to successfully process audio this way.

### 2.1.3.3 MIDI Tracks

Running as a track plug-in, the TimewARP 2600 can process prerecorded MIDI tracks.

To do this, set up two tracks just as you did above in section 2.1.3.1: an audio track that is running the TimewARP 2600 as a track insert, and a MIDI track whose output is routed to the TimewARP 2600.

When you play this track (use the Pro Tools transport window to play, rewind, pause the MIDI sequence, etc.), the MIDI events stored in the track sequence will be routed to the TimewARP 2600. MIDI note-events will become key depressions at the virtual keyboard; MIDI pitch-bend will turn the virtual keyboard pitch-bend knob; and MIDI controllers that are mapped to TimewARP 2600 sliders, knobs or switches (see section 2.2) will move those controls.

### 2.1.4 Selecting Patches

The TimewARP 2600 gives you a three-level hierarchy for storing and organizing your patches. All *Patches* are sorted into various *Categories*, which are in turn sorted into major *Groups*. Each of the three patch selection buttons generates a drop-down list associated with one layer in this hierarchy.

*Groups*, *Categories*, and *Patches* can also be selected by keyboard shortcuts. *The up/down arrow keys* on the computer keyboard select *Patches*, the *left/right arrow keys* move between *Categories*, and using the *control key* with the *left/right arrow keys* moves between *Groups*.

### 2.1.5 Making Patch Connections

Use the mouse to connect any output to any input jack. Position the cursor at any signal output, click and hold down the mouse button, and drag the patch cord to any signal input. To remove a patch cord, drag either end away from its signal jack.

The TimewARP 2600 will not connect two outputs or two inputs together. If you drag from one jack to another but no patch cord appears, it may be that both jacks are inputs or both are outputs.

The TimewARP 2600 will allow Y-connections, distributing a single output signal to multiple destinations by holding the control key down while clicking an output that already has a cable connected to it. A second cable will appear attached to the cursor, which can then be plugged into another input.

### 2.1.6 Using the Sliders and Knobs

To adjust a slider or knob, drag it with the mouse. For increased resolution, hold down the *command* (Apple) or *control* (Windows) key during the drag operation.

If you pause for a moment over a slider, a display will pop up with the numerical value and name of the parameter it controls.

If you have MIDI controller hardware, you can easily connect it to the sliders, knobs or switches; see section 2.2.2.

### 2.1.7 Using the Virtual Keyboard Display

The graphic TimewARP 2600 keyboard display responds directly to mouse events; click on any key to create a MIDI keydown signal. This is useful when you are creating and tuning a new patch, if you don't have a hardware MIDI keyboard handy.

The virtual keyboard also displays – as key depressions - incoming MIDI note-events. Use this to monitor your external MIDI keyboard connection and activity.

### 2.1.8 Polyphonic Operation

The TimewARP 2600 can respond to as many as eight simultaneous keydown events. Use the *Voices* dropdown (see section 4.1.1.2) to set the number of voices.

### 2.1.9 Saving and Loading Patches

You may save patches using the *Save* or *Save As* buttons or load patches with the *Patch Manager* (see section 4.1 below). There is no limit to the number of patches you can create and save. To enable the *Save* button, the *Lock* button (padlock icon) must be in the unlocked mode.

You may wish to create a bank of “template” patches, generic versions of often-used configurations.

The TimewARP 2600 responds to MIDI bank-select and patch-select commands.

## 2.2 MIDI Communication and Control

### 2.2.1 Slider, Knob and Switch Control

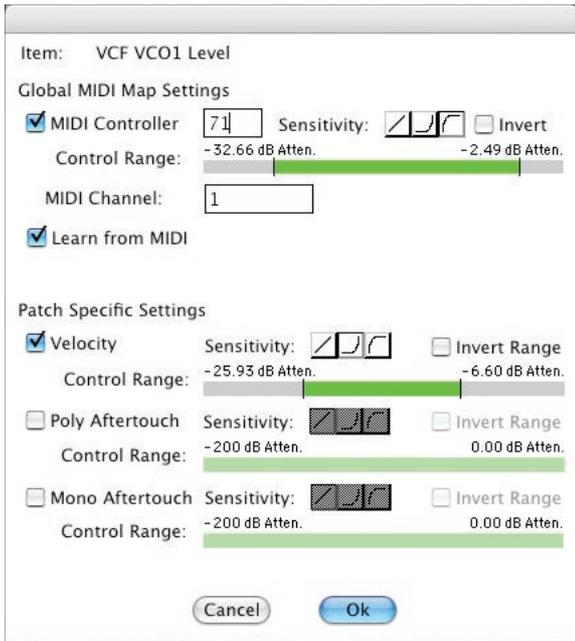
You may control any slider, knob or switch on the TimewARP 2600 plug-in with any MIDI controller. If you have a hardware control surface, or if your MIDI keyboard has slider or knob control devices, you can use them to control any combination of sliders, knobs or switches in a patch.

Any connections you set up are global to the TimewARP 2600 – you can store them independently of any specific patch (see section 4.1.2.1 below).

To connect a slider, knob, or switch to an external MIDI controller, hold down the *control*

(Mac) or *shift* (Windows) key and click on the slider, knob or switch image. The responding dialog box offers you both global and patch-specific connections.

### 2.2.1.1 Global MIDI Map Settings



Select the *controller number*, either by directly typing it in, or by twiddling the physical control on the MIDI device you intend to use.

Set the *response curve*, *polarity*, and *range*. The response curve may be linear, or exponential, or logarithmic. The polarity is either direct or inverted. To set the *Control Range*, drag either end of the bar inward. Whenever the bar covers less than the entire controller range, you may also drag it to the right or left, to adjust the range offset.

You may, if you wish, assign the same external control to several TimewARP 2600 sliders or knobs, and set a different curve, polarity, and range for each one. In this way, you can create patches and configurations of the TimewARP 2600 that are specifically adapted for expressive performance either live or in the studio.

### 2.2.1.2 Patch-Specific Settings

Besides globally assigning a slider, knob or switch to a MIDI controller, you may also assign the slider or knob to be controlled by the velocity and/or aftertouch parameters of incoming keyboard events. (These assignments are *not* global; they are stored with individual patches.)

In real time, these controller values are summed with the current global control values to determine the real-time value of the slider. So, by careful tuning beforehand, you can have both keyboard expression and hardware knob/slider control at the same time.

## 2.2.2 MIDI Patch Selection

In each group of patches, the first 128 are available to standard incoming MIDI patch selection commands.

Patch categories don't affect this numbering. For example, if the first category in a group has 127 patches, then the second category will have just one patch available for MIDI patch selection.

Groups themselves can be selected with MIDI bank-select commands



# The Craft of Audio Synthesis 3

This chapter is about the facts – physical, mathematical, and auditory – that make the TimewARP 2600, and the hardware that it emulates, possible. We have to spend a few minutes here distinguishing between physical signals, and the sounds that people hear in the presence of certain kinds of signals.

This is important because synthesizer equipment can only deal with signals – physical commotion of one sort and another. When you are fiddling with synthesizer equipment, you are generating and modifying signals for the sake of the interesting (we hope) sounds you hear when those signals reach your eardrums.

## 3.1 Signals and Sounds

A signal is something happening: a waving flashlight, ambulance siren, referee flag dropping, winking at a friend. Tiny disturbances of the air around us are signals for our ears; we hear them as noise, or singing, or sirens, shrieks, growls, whatever.

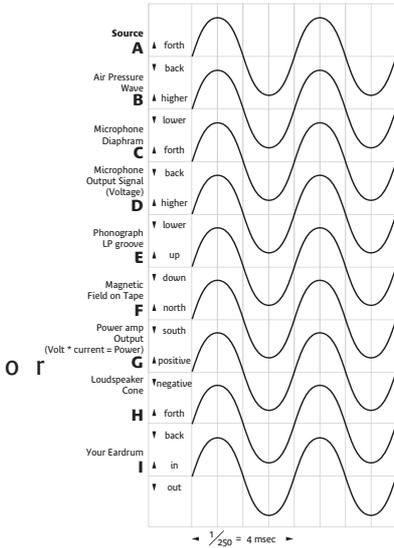
The *signal* is the physical disturbance in the air, the movement of the eardrum. The *sound* is your perception of the signal: “Hello!”

### 3.1.1 Analog and Digital Representations of Signals

The signals we are concerned with in sound synthesis are *audio* signals: more or less regular variations in air pressure, at our eardrums, repeating at rates of between 20 and 20,000 times in one second.

Such signals are straightforward physical processes which can be recorded and reproduced. One way to do that is to look at the pattern of air-pressure variation, and model it in some other medium. During the past century this has been done with grooves in a phonograph record, magnetic fields along a length of tape or wire, and other media. The usual scenario is: with one or more microphones, generate an electronic model of the vibrating air, then use the electronic signal to drive a magnetic recording head, or amplifier, or LP recording lathe. Throughout such processing, the signals we deal with are directly analogous to each other; except for the change in medium from air to voltage to magnetic field strength or stylus position, the signals are identical. Graphed or charted, they even *look* the same. This is *analog* recording.

Analog signals @ 250 Hz  
(approximately middle C)



Another way to record such a signal is, with high-speed digital circuits, to measure the underlying medium many times each second, and store the measured numbers. This is *digital* recording.

Since the TimewARP 2600 is not constructed of electronic circuits, concepts such as “voltage” don’t apply here. The TimewARP 2600 is software, a complex piece of computer behavior. The signal medium for the original ARP 2600 was electrical pressure, measured in volts. The signal medium for the TimewARP 2600 is simply number sequences. We use those numbers the way the original machine used electrical pressure; where the original Owner’s Manual used the word “voltage” we will just say “signal” “signal level”, and in the module specifications we will refer to “virtual Volts” or “vV”.

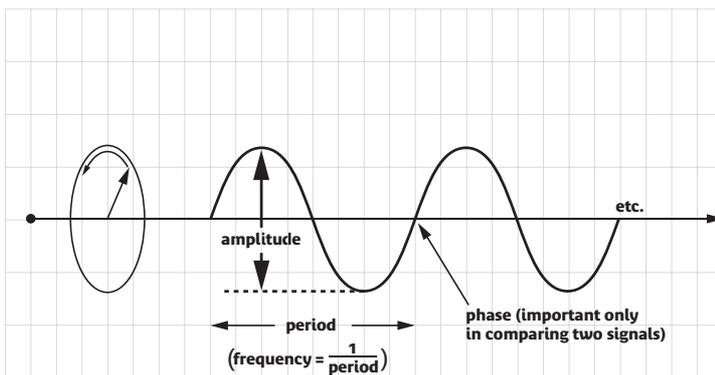
### 3.2 Attributes of Signals

The simplest possible signal is a sine wave. It’s like the back-and-forth motion of a point on a circle as the circle rotates. A lot of the mathematics of sine waves is based on that rotating-circle idea; you don’t have to get involved in that unless you’re curious about it, but it’s helpful to train your imagination by picturing the basic sine-wave graph as a slightly stretched coil spring like a “slinky”.

The motion of a pendulum, or of a tuning fork, swinging back and forth as they slowly come to rest, is a *decaying* sine wave.

A sine wave signal has exactly three attributes: its *frequency*, its *amplitude*, and its *phase*. It has no other characteristics at all. (The decaying swing of the pendulum or the tuning fork does have one other attribute: the amount of energy/amplitude that it loses on each swing. It’s not a “pure” sine wave.)

#### 3.2.1 Fundamental Attributes



Picture a point on that rotating circle, leaving a trail behind as it rotates, like an airplane propeller. Picture the trail, stretched out behind like a coil spring, and ask yourself:

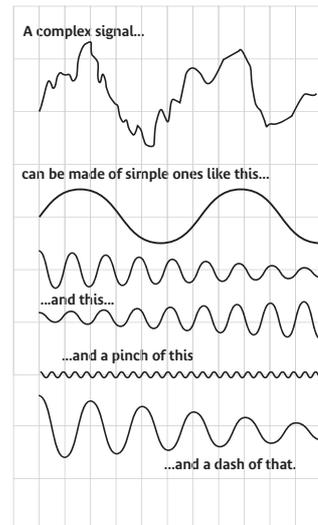
**3.2.1.1 Amplitude: what's the diameter of this imaginary circle?****3.2.1.2 Frequency: how fast is it rotating?****3.2.1.3 Phase: when does it start a new cycle?****3.2.2 Complex Attributes**

Most of the activity that reaches our ears every day is far more complex than just a sine wave. Banging on a garbage can creates a *much* more complicated sort of vibration than tapping a tuning fork.

But any kind of motion that *isn't* a sine wave can still be analyzed as a collection of sine waves. This is called *Fourier analysis*, after the man who discovered that it was possible, and proved – mathematically – that it always worked.

This goes both ways: any complex signal can also be *constructed* from a collection of sine waves with the appropriate attributes: amplitude, frequency, and phase relationships.

In fact, *any kind of motion* – any repeating pattern of the sort that we are interested in here as “sound waves” - can be examined and analyzed under two different headings:

**3.2.2.1 Waveshape, and Time-Domain Attributes**

Any repeating motion can be diagrammed in a graph where the horizontal axis is a period of time and the vertical one is the motion itself, back and forth. This is *the time-domain* view of a signal:

**3.2.2.2 Spectrum, and Frequency-Domain Attributes**

Instead of looking at a signal as something in motion through a period of time, we can look at the collection of sine-wave components of the signal. In such graphs, the horizontal axis is a frequency range (instead of a time period), and we indicate each spectrum component with a single vertical line in the graph. The height of the line indicates the strength of the component at that frequency. This is the *frequency-domain*, or spectral-domain, view of a signal.

It is the distribution and relative strength of spectral components that we experience as the tone-color of a sound or sounds. “Bright”, “dull”, “sharp”, “tinny”, “heavy”, and so on – these are all descriptive words for the spectral attributes of sounds.

3.2.2.2.1 Harmonic Series

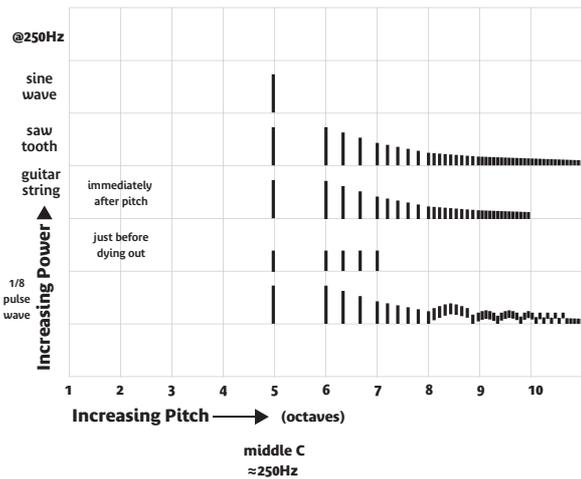
The spectral view of any periodic signal has components at simple multiples of the signal frequency. For example, suppose we examine a sawtooth wave at a frequency of 110Hz. It will have components at 110, 220, 330, 440, and so on.

A simple number sequence like this is called a *harmonic series*. It is interesting to think about musically; the smaller numbers in the series all form simple musical intervals: octaves, fifths, fourths, and thirds.

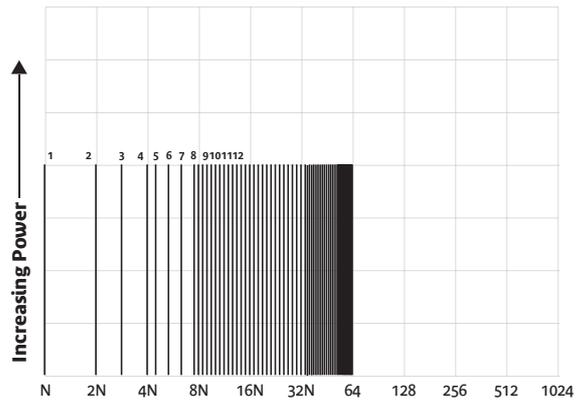
Most musical instruments generate harmonic spectra. Some have more, some have fewer harmonics, and there are wide spectral variations even within a single instrument depending on how it is played. In general, whatever the actual spectral components are, they will always form a harmonic series.

In audio synthesis, you will use oscillators to generate pitched tones that have a harmonic spectrum.

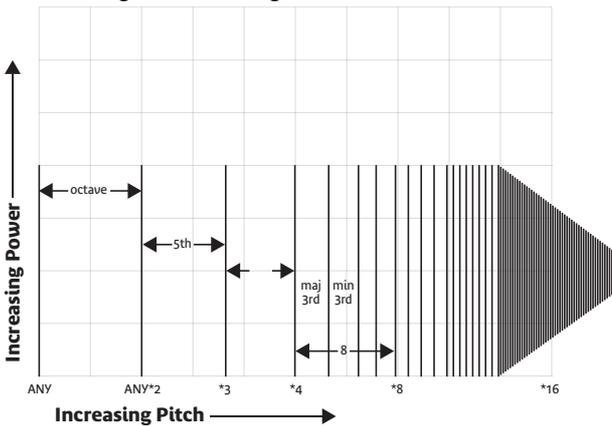
Spectrum = the sine-wave component of a signal



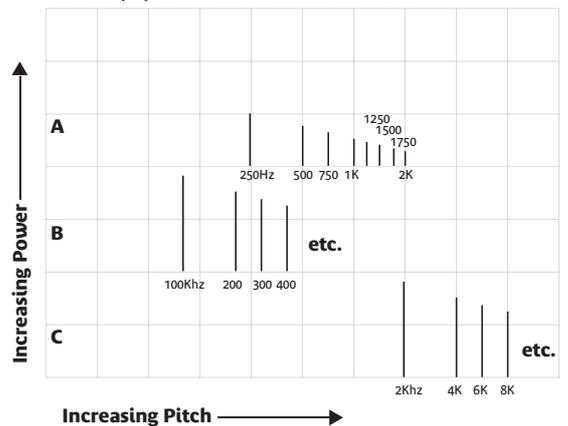
A harmonic series is composed of numerically equal frequency intervals, and that means decreasing pitch intervals



The first couple of octaves' worth of a harmonic series make useful musical intervals: octaves, 5ths, 4ths, 3rds and so on. Beyond the 16th, they get too close together to sound good.



Pick any frequency; the multiples of that are a harmonic series. A, B, and C are all harmonic series.

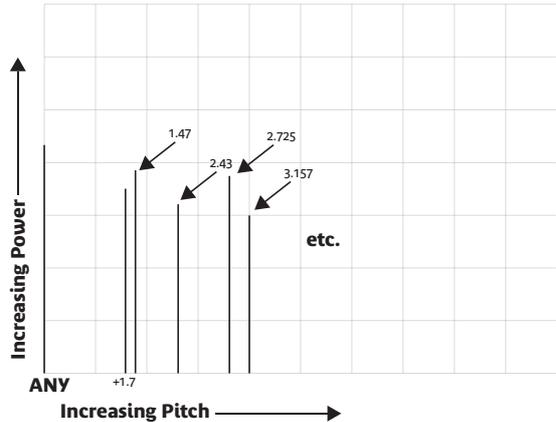


**3.2.2.2.2 Enharmonic Series**

Enharmonic, in this context, means “not forming a harmonic series.” Some waveforms do not repeat themselves at a regular interval; the spectra of such waves will have components at strange non-integral frequency ratios, or they may have shifting spectral components that die out and reappear. Some percussion instruments, such as kettledrums, or bells, have enharmonic spectral components.

In audio synthesis, you can use various modulation techniques – such as AM and FM - to generate enharmonic spectra.

Not all spectra are harmonic. An Arbitrary collection of sines at unrelated frequencies is “unharmonic”.



**3.2.3 Attributes of Random Signals**

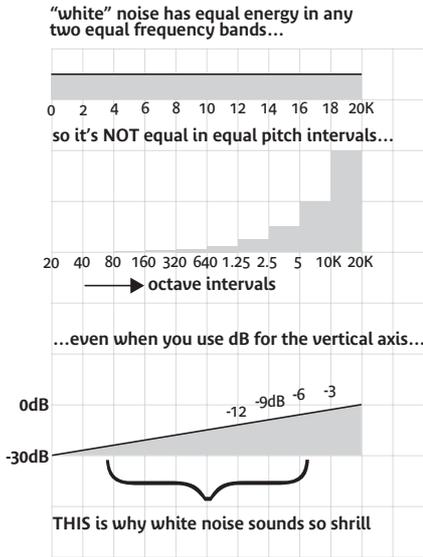
A completely random signal – noise – actually has a *continuous* spectrum: it does not consist of isolated sines in a harmonic series, nor even in an enharmonic series. A noise spectrum is a continuum of frequency components; in order to describe it, we have to talk about how much energy is in each band in this continuum. One kind of noise may have high-frequency energy, and another has low-frequency energy.

In audio synthesis, noise is an extraordinarily useful signal. Filters can be used to shape a noise spectrum into almost anything – even pitched sounds.

**3.2.3.1 Spectral Balance (Color)**

The physical processes (such as molecular motion or random popping of electrons from one atom to another) that we typically depend on for random electronic noise have a frequency distribution of equal probability at any frequency, or through any frequency interval. Curiously enough, this produces noise signals that, for audio purposes, don’t sound properly balanced across our range of hearing. They sound too bright.

### 3.3 Attributes of Auditory Events



Of all the thousands of sounds you hear in a single day, only some have a definite beginning or ending point. Many just come and go without a clear start/stop. Those that do have a reasonably clear start/stop we will call auditory events. The other sounds, the less clearly defined ones, we can call auditory *textures* or *backgrounds* or even *structures*.

Think how some sounds are more complicated than others. The hum of a refrigerator is sort of simple, the hum of a tuning fork is (as we heard above) really simple. A single bass note from a piano is quite complicated, the way it keeps changing and evolving as it dies away. Human speech is a very complicated kind of sound, even when it's your native tongue (when it's a language you don't understand, it sounds even more complicated).

There are many, many more kinds of sounds to hear than you or I have words to describe them with. Here are just a couple of the most general distinctions people make about different kinds of sounds:

**3.3.1 Steady-State Attributes:** some sounds, once they get started, remain pretty constant. They don't change much while they continue, and when they stop, they just stop. We say such sounds reach a *steady state*, in which we can pick out a definite *Pitch*, *Loudness*, and *Tone Color*. Notes from organ pipes – or from electronic drawbar organs – are steady-state sounds.

**3.3.2 Time-varying Attributes (i.e. "Envelopes"):** some sounds have a pretty clear start and end, but while they're happening they change. Think of a bird song, or ordinary human speech.

Such sounds can sometimes be analyzed as more or less rapidly changing in one or more of those fundamental auditory attributes: pitch, loudness, or tone-color. Think, for example, of how the sound of a plucked guitar string evolves from the moment you pick it to the moment you can't hear it anymore. It starts out loud and fades away; and it starts out "bright" - with lots of harmonics – and is slowly "muffled" as it fades out.

One way to conveniently diagram what happens in such events is to chart each changing attribute separately, in its own time-domain graph.

Such a graph is often referred to as an *envelope*. Over the years, some standard vocabulary has developed for talking about envelopes: *Attack* (for the first part of an event), *Decay* (for some later parts), and *Release* (for the last part, when the event is coming to an end).

### 3.4 How Signals and Sounds Go Together... Sort Of

Signal activities, being entirely physical kinds of things, are easily nameable, measurable, catalogable, countable. Sounds, as we pointed out above, are not quite so easily domesticated. The music and other sounds that we listen to correlate in several well-established ways with the signals around us; but the correspondence is not simple. There are always surprises.

#### 3.4.1 Signal Frequency and Audible Pitch

This is probably the best-established and most reliable correspondence. It dates all the way back to Pythagoras, the Greek philosopher who 2500 years ago worked out that halving the length of a vibrating string made the pitch rise by one octave.

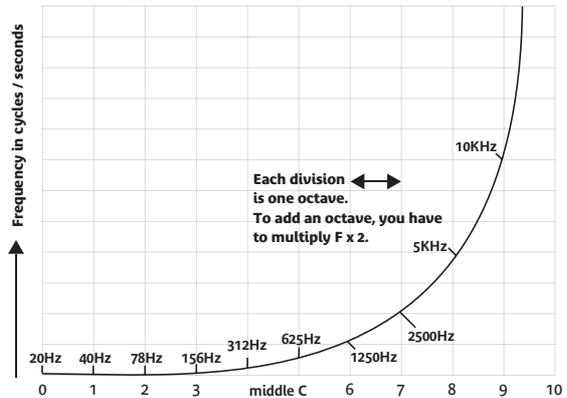
##### 3.4.1.1 How Adding Pitches Means Multiplying Frequencies

The range of audio frequencies – of human hearing, in other words – is conventionally stated to be 20Hz to 20KHz. And when you think linearly, it sounds tragic to learn of someone, say, who can only hear up to 10KHz. But in the realm of human pitch perception, such a person has kept 90% of his hearing range: from 10KHz to 20KHz is just one octave out of the 10 we hear.

This is a *really important fact* in audio synthesis; you will encounter it over and over again, in every patch you create. It governs much of the arithmetic of generating and controlling spectral distributions and patches:

- ▶ Equal *pitch* intervals require exponentially increasing frequency increments;
- ▶ Equal *frequency* increments make a harmonic series. In a harmonic series, as we go up the series we get smaller and smaller pitch intervals.

**A harmonic series is composed of numerically equal frequency intervals, and that means decreasing pitch intervals**



### 3.4.2 Signal Amplitude and Audible Volume

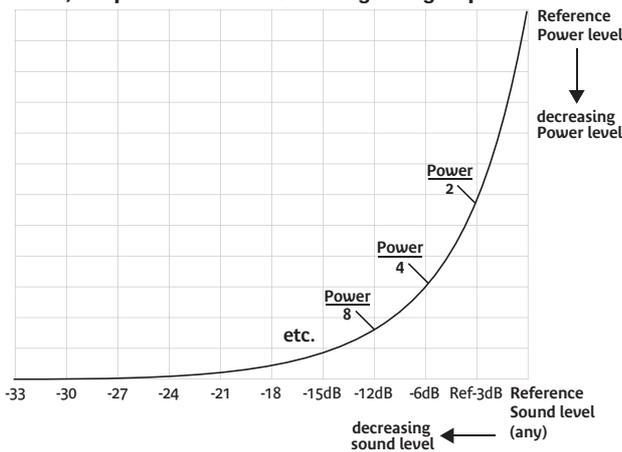
This is a very dicey relationship. It is true that, for any given signal, increasing its amplitude will increase the volume of the associated sound. But a lot of other signal characteristics have a greater impact on our perception of volume than amplitude does.

For example, the difference between talking and shouting at someone is far more a matter of pitch and spectrum – tone-color – than of mere amplitude. When I yell, I “raise my voice”; that is, I raise the pitch of my voice, and I put more energy into it, which generates more harmonics, which is a matter of spectral content, not amplitude. Stage actors have to learn to overcome this tendency; in order to be heard onstage without shouting, they have to learn to increase their speaking amplitude without yelling.

If you examine – visually - the recorded signal from a pipe organ, or even an orchestra, you may be surprised to find the apparent amplitude of the softer signals almost equal to that of the loudest. This has to do with the spectral distribution of the sound. Of two different signals of approximately equal amplitude, the one with the broadest spectral distribution – the most harmonic content - will sound louder.

#### 3.4.2.1 How Adding Volumes Means Multiplying Amplitudes

A 3dB rise or fall in sound level is noticeable, but it's not much; it represents a factor of 2 change in signal power.



Nonetheless, for a given signal and its spectrum, it is always true that a bare increase in amplitude will be heard as an increase in loudness. But how much of an increase?

It turns out that, just as with frequency and pitch, the relationship here is *exponential*. A century of research has established that equal steps in loudness are represented by multiplicative ratios in signal amplitude.

In other words, if you double the amplitude of a signal, you will hear an increase in loudness. But then to get another increase equal to the first one, you have to double the amplitude again.

### 3.4.3 Signal Spectrum and Audible Tone-Color

This is a quite solid relationship. In fact, the word “spectrum” has achieved a meaning in both worlds: depending on the context, it can refer to a measurable attribute of physical signals, or a character of perceived sounds.

Any change you hear in the character of a sound – in its tone-color or “spectrum” - must have an associated variation in the character of the physical signal that is arriving at your eardrums. Tone change = waveform change.

The reverse is not quite so certain. It’s fairly easy to find waveform changes that listeners can’t hear. For example, we humans simply aren’t sensitive to phase relationships within a complex spectrum. But, in the time domain, two identical spectra with shifted phase relationships among their components can look unrecognizably different.

### 3.4.4 Signal Envelopes and Audible Event-Contours

One of the most fascinating areas of audio synthesis is listening for the envelopes of time-varying events. Here there are all sorts of mysteries, in which signal attributes get regularly “misperceived”: frequency variations get heard as volume, spectral evolutions are heard as pitch, and amplitude envelopes generate an elusive spectral twitter.

### 3.5 Modules and Methods for Generating Signals

How can you get access to the signals and processes we’ve just described, so that you can play with them on your own?

Throughout the past century people have been noodling around with electronic ways of generating audio signals. In particular, beginning in the 1960’s, people such as Bob Moog, Don Buchla, and Alan R. Pearlman began to settle on some ideas that have become almost standard for audio synthesis: *independent, modular* functions for signal generation and signal processing, capable of being controlled not only by hand but also by signals of the same kind as they generate.

This was the idea of *voltage-controlled* operation, and it was completely revolutionary. Using independent, modular functions made it possible to change one attribute of a signal without necessarily affecting any other attribute; so the craft of synthesis became the craft of constructing, and tuning, integrated configurations of modules.

The cables that connected modules were called *patchcords*, and so connecting modules together came to be referred to as *patching* them, and so, finally, any working configuration came to be called a *patch*.

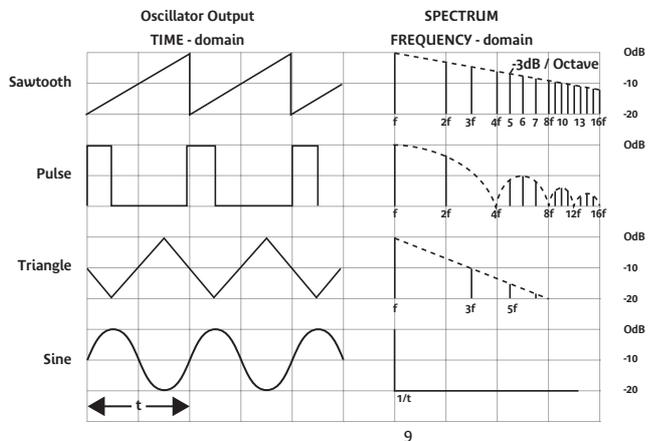
Let’s take a look at some modules that generate signals.

#### 3.5.1 Oscillators

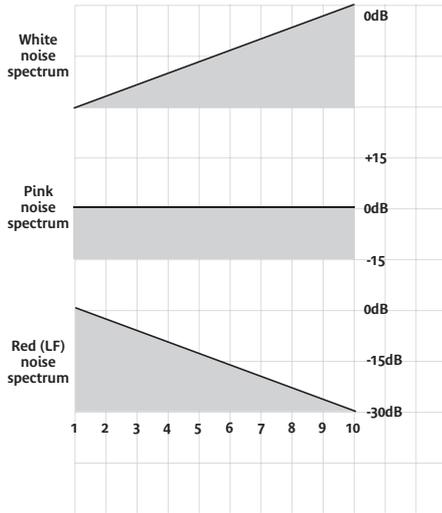
A device that repeats the same motion over and over is an *oscillator*. In audio synthesis, oscillators typically produce very simple geometrical-pattern signals such as sine, triangle, pulse, and sawtooth waves, named simply for what their time-domain graphs look like.

In an *analog* synthesizer, the underlying medium in motion is usually electrical pressure, or voltage. In a *digital* synthesizer, the signal is actually generated as a sequence of calculated numbers. (Conventionally, these are generated at the standard sampling rate for music CDs, 44.1KHz.) This does not become motion until, at the output of the synthesizer, the number stream is converted to variations in voltage, and then amplified, and then used to drive a loudspeaker. (A loudspeaker is a motor that moves back and forth instead of around and around.)

Because oscillator-generated signals are periodic, their spectral components always form a harmonic series.



### 3.5.2 Noise Generators

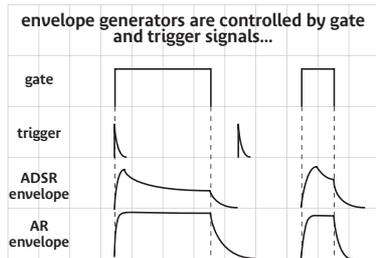


A device that jiggles at random without ever repeating itself is a *noise generator*. Waterfalls, steam, wind, fans, and such things are all noise generators.

The spectrum of a noise signal is a statistical distribution of frequency components. (This is the opposite of a sine wave, which is exactly one frequency.) A noise spectrum that is perfectly balanced throughout the musical range is called *pink noise*. Pink noise is very useful in listening tests of loudspeakers, because a trained human listener can hear even tiny differences between two different noise spectra.

Filtering and equalization can shape a noise spectrum into almost any sound.

### 3.5.3 Envelope Generators



A device whose output is intended to control some time-varying attribute of an event is called an *envelope generator*. These are sometimes referred to as *transient generators*, to call attention to the fact that their output is not constant but transient.

An envelope generator produces an output signal only “on demand”. The demand is made by means of timing signals called *gates*, and *triggers*.

### 3.5.4 Sample & Hold Processors

The idea of “sampling” a signal does not directly relate to any particular characteristic of audio events; instead, it is an idea from electronics that has turned out to be useful for creating patterned control signals.

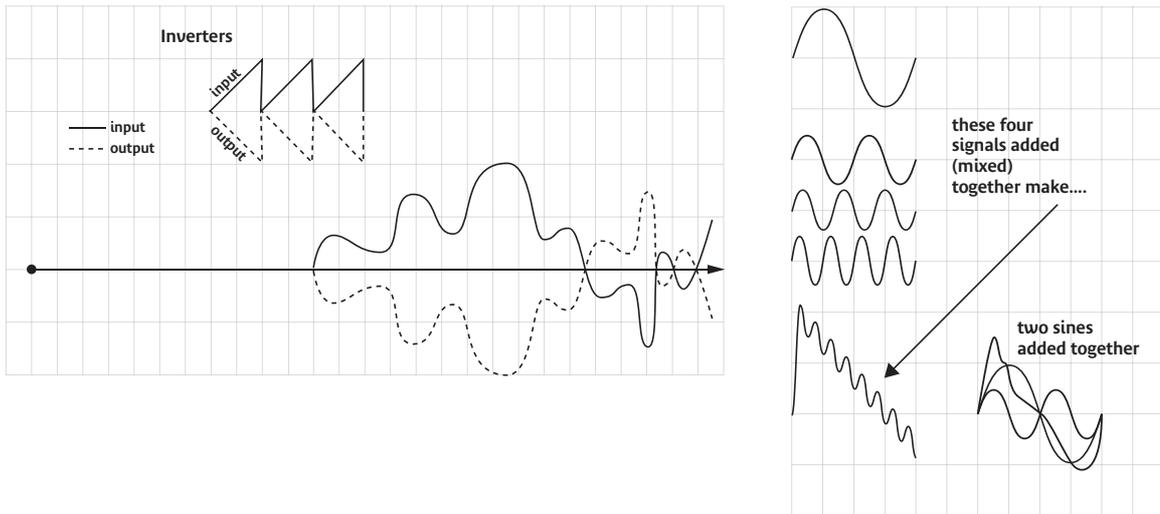
## 3.6 Modules and Methods for Processing/Modifying Signals

### 3.6.1 Inverters

A signal inverter works exactly like a seesaw. When the input goes high, the output goes low, and vice versa. An analog inverter would output negative voltages on positive input; a digital inverter simply multiplies its input number-stream by (-1).

### 3.6.2 Signal Mixing

A signal mixer adds two or more signals together and outputs the result of the addition. This is a more complex signal, usually, than any of the inputs. But not necessarily; if signal B, for example, is the exact inversion of signal A, then mixing the two will produce a signal of exactly zero.



### 3.6.3 Attenuators/Amplifiers

An attenuator cuts the strength of a signal passing through. Digitally, this is accomplished by multiplying the input by some value ranging from 0.0 (which passes no signal) to 1.0 (which passes the signal at its full input amplitude).

An amplifier may increase the amplitude of a signal. But not necessarily; it is usual for a voltage-controlled amplifier to have a maximum gain factor of 1.0. In fact, the purpose of VCA's is actually to "chop" the signals passing through them. It would make more sense to think of them as "voltage controlled attenuators".

#### 3.6.3.1 Gain Factor

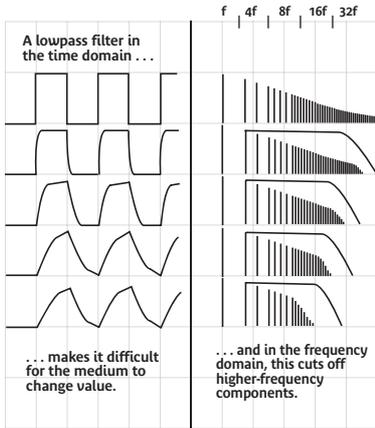
To describe the behavior of an amplifier or attenuator, we may use the expression "gain factor" to mean the ratio of output signal amplitude to input amplitude.

### 3.6.4 Filters

A filter is a device that works better at some frequencies than at others. (The inverters, mixers, and attenuators we have been describing work the same at all frequencies, so they are not filters.)

Because of this frequency-dependent characteristic of filters, they change the shape of any complex waveform passing through. And so it will be important for you to get to know what filters do to signals in both the time domain and in the frequency domain.

### 3.6.4.1 Low-Pass Filters



Any device or mechanism that passes along slower motions better than faster ones can act as a lowpass filter.

Picture yourself stirring a cup of tea with one of those little wooden paddles they hand out in the coffee shops. Stir it back and forth, fast. Now slow down. Now imagine the tea has turned to syrup. You can still stir it slowly, but if you try to go fast the stick will simply not move.

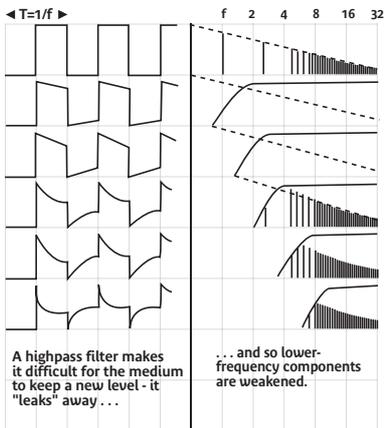
That's a lowpass filter. You can see the effect of this on a signal quite easily.

For audio signals, you will usually be more interested in the frequency-domain effects of filtering. For subaudio signals, it is usually the time-domain effects – changes in waveshape – that we care about. In the time domain, a low-pass filter rounds off

any sharp transitions in the signal. A good example of this is the *lag processor* described in section 3.6.4.5 below.

In the frequency domain, it weakens spectral components that are higher than the filter *cutoff frequency* – the frequency at which the filter begins to have an effect on the signal.

### 3.6.4.2 High-Pass Filters



Any device or mechanism that passes along faster motions better than slower ones is a highpass filter.

Take the drinking straw from your water glass. Seal the end of it with your thumb and dip it back into the water. Notice that you can pump it up and down in the glass, fast or slow, but the water never leaks into the straw as long as you hold your thumb over the end. Think of the up and down motion of the water at the bottom end of the straw as “the signal”.

Now start letting a little air leak into the straw as you move it up and down. The water level at the bottom end of the straw no longer stays down when the straw “signal” goes down – it starts to come up again. And then when you draw the straw back up, the water leaks back down more or less rapidly depending on the position of your thumb at the top of the straw.

This is a high pass filter. In the time domain, it constantly “leaks” its output signal level back to zero, at a rate related to the cutoff frequency and slope. In the frequency domain, it passes all spectral components higher than the cutoff frequency, and attenuates those below the cutoff frequency by an amount proportional to the cutoff slope.

### 3.6.4.3 Cutoff Slope

This is the rate at which a filter attenuates spectral components, as a function of their frequency. It is usually a multiple of 6dB/octave.

### 3.6.4.4 Feedback and Resonance

It is usually possible, with any signal-processing device or system of devices, to mix some of its output signal back into the input signal. This may be intentional, or it can happen by accident; everybody who has ever worked with a PA system has experienced the terrible screech of a system “in feedback”.

In filter modules for audio synthesis, it is common to provide controllable feedback. With just a little, the filter response begins to peak around its cutoff frequency; as the feedback level increases, the peak gets stronger. Eventually – just as happens with a PA system – the filter falls into *oscillation*. Regardless of the input signal, it “screams” a sine-wave at its cutoff frequency. In this state it is no longer behaving as a filter at all; it has become an oscillator.

Systems in feedback have been very well studied in physics. Their behavior can be described mathematically. Whereas feedback in a PA system can be unpredictable and uncontrollable, feedback in a filter module for audio synthesis can be – and is – a controllable and useful feature.

### 3.6.4.5 Lag Processors

A *lag processor* is a low-pass filter intended specifically for processing subaudio signals. It introduces a “lag” in the output signal wherever the input shows a sharp change in value. How much of a lag depends on the “cutoff frequency” of the processor.

## 3.6.5 Modulation Methods

The simplest possible signal, we said above, has exactly three attributes and no more: frequency, amplitude, and phase. If, starting from a steady-state signal, we systematically modify any of these characteristics, we are said to be modulating the signal. And so, based on these three signal attributes, there are three possible forms of modulation: Amplitude Modulation (AM), Frequency Modulation (FM), and Phase Modulation (PM). The first two are more commonly used in audio synthesis than the third; we won’t say anything here about phase modulation.

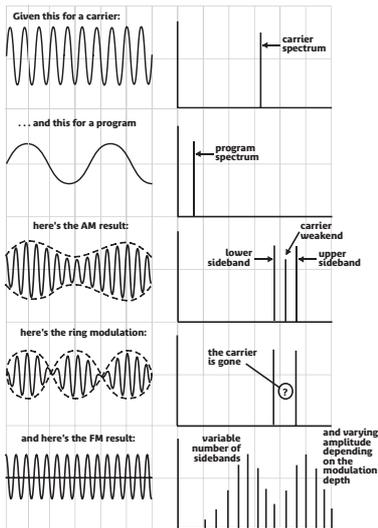
Using AM and FM methods, it is possible to generate waveforms and spectra that are far more complicated – and interesting to the ear – than anything that can be produced by merely mixing and filtering signals. Mainly that is because of sidebands.

### 3.6.5.1 Sidebands and Sideband Spectra

What happens to the spectrum of a sine wave when we modulate its amplitude? What happens when we modulate its frequency?

Clearly, since the signal that results from AM or FM methods is no longer a plain vanilla sine wave, then, in the frequency domain, it must have some additional components. These additional components are called *sidebands*. They have been studied for at least a century, and are pretty well understood physically and mathematically.

Audio synthesis is probably the only application of AM or FM modulation where we are interested in sidebands for their own sake, as something to listen to directly; in the past, this stuff was only interesting to radio engineers, radar, sonar, television broadcast engineering. In those disciplines, modulation sidebands are products of broadcasting methods, in the electromagnetic spectrum. Because of that historical background, some of the conventional language for talking about modulation processes is a little weird: the signal being modulated is often referred to as the *carrier signal*, and the signal that provides the modulation pattern is called the *program*. (Guess why.)



Any form of modulation generates, for each component of the original signal, at least one *lower sideband* – at a frequency equal to the component minus the modulating frequency – and at least one *upper sideband*, at a frequency equal to the component plus the modulating frequency. If the carrier – the original signal - is itself complex, with multiple spectral components, then each of its components will produce its own sidebands independently of all the others. Likewise, if the modulating signal – the program - is complex, the arithmetic applies separately to each of its components.

So, just for example, if you modulate a 10-component carrier signal with a 10-component program signal, the signal resulting from the modulation will have not less than 100 spectral components. This can get very messy; the most useful thing you can do with such a signal, before you do anything else, is filter it to get rid of some of the fuzz.

### 3.6.5.2 Amplitude Modulation

Suppose we modulate the amplitude of a 1000Hz sine wave with a 5Hz sine wave. The result is indistinguishable from what we would get if we mixed three sine waves, at 995Hz, 1KHz, and 1005Hz. They are the same signal.

The 995Hz component of the output is the lower sideband resulting from the modulation, and the 1005Hz component is the upper sideband.

### 3.6.5.2.1 Ring Modulation

Whereas a VCA responds only to a positive-going signal at its amplitude-control input, a ring modulator responds to both positive and negative levels at both of its inputs. Its output is simply the product, arithmetically, of the two inputs. If you are new to audio synthesis, draw a couple of signal graphs – it doesn't matter what they are – on the same timebase and vertical scale, and use a pocket calculator to work out the result of multiplying the two signals together. That's what a ring modulator does. (The expression "ring modulator" describes the appearance of the analogue circuit design that's required for the multiplication.)

In the frequency domain, the difference between this and ordinary AM is only that the carrier signal components are suppressed. Once again, if you work out the arithmetic, a single-frequency carrier, modulated by a single-component program, generates a three-component AM spectrum but only a two-component ring-modulation spectrum.

What's useful about this? Well, since the carrier is almost always periodic (it comes from an oscillator, right?), it has a harmonic spectrum. *Suppressing* this spectrum lets you hear just the sidebands, which can be completely enharmonic if you're careful about the ratio of the two input signal frequencies.

### 3.6.5.2.2 Frequency Shifting

It's theoretically possible not only to suppress the original carrier, as in ring modulation, but to isolate the lower and upper sidebands and make them available separately. The arithmetic here is fascinating, because the end result (for once) is in one-to-one correspondence with the input: for each component of the program signal, there is a component in the output at C-p (or at C+p). In other words, the final spectrum has only as many components as the original program did. This is called frequency shifting. Picture the entire program signal spectrum shifted up or down by some fixed frequency.

The important thing to remember about this is that it's not pitch shifting – which would have to be accomplished by frequency multiplication – but *frequency* shifting. It's an addition or subtraction process, and it *really* messes up any harmonic relationships that might have existed in the original spectrum.

### 3.6.5.3 Frequency Modulation

The spectrum resulting from amplitude modulation always has three components for every one component of the program signal: the carrier itself, and two sidebands. In *Frequency Modulation*, however, the number of sidebands depends on the modulation depth. It is possible from only two sine waves to generate a spectrum with dozens or even hundreds of components. Modulating one sawtooth with another can produce a spectrum so complex that it sounds almost like a noise generator. In such a patch, you will usually reach for a filter to take the edge off the resulting spectrum.

What happens is this: as the depth of modulation increases, the number of sidebands does too, without limit. The additional sidebands come in at – guess what – integral multiples of the program frequency.

For this reason, the most useful FM techniques involve only sine-wave carrier and program signals.

### 3.7 Controlling One Module by Means of Another

The modulation methods we've just described can be accomplished with voltage-controlled or digitally-controlled equipment. For example, to set up an AM effect, feed the carrier signal into a VCA audio input, and the program signal into one of the amplitude-control inputs.

Likewise, to set up an FM method, route the program signal into one of the frequency-control inputs of a VCO. (See section 4.3 for news of some TimewARP 2600 extensions relating to this.)

#### 3.7.1 Linear and Exponential Sensitivity To Control Signals

In designing and constructing a voltage-controlled device, or a digitally-controlled algorithm, we have to consider how we intend the controlling parameter to relate to the controlled parameter.

Suppose, for example, we build an oscillator that changes its output frequency by exactly 1000Hz for each rise of 1 volt in electrical pressure at a designated point (a "control input") in the oscillator circuitry. This is a linear relationship between the control value and the frequency. In order to march such an oscillator through a musical scale, we would have to provide exponentially increasing voltage steps for each rising musical interval. That's pretty awkward, and it gets worse very rapidly. The pitch interval produced by, say, a 0.5-volt control step will depend – completely – on the initial frequency of the oscillator.

Suppose we tune this imaginary linear-responding VCO to 500Hz, and apply a sequence of 0.5V control steps. What frequency do we get at each step? What musical interval?

INITIAL CV	FREQUENCY	INTERVAL
0	500Hz	unison
+0.5V	1KHz	octave
+1.0V	1.5KHz	Fifth
+1.5V	2.0KHz	fourth
+2.0V	2.5KHz	Major third
+2.5V	3.0KHz	Minor third

And so, instead of linear sensitivity, VCO's for audio synthesis are designed, almost universally, with *exponential* sensitivity to control signals. Such a design sets up an exponential relationship between control signal and frequency *in order to maintain a simple linear relation between control signal values and audible pitch changes.*

Similar arguments apply in the design of voltage-controlled filters: it's more practical to work with a linear relationship between control-signal and "cutoff timbre", and therefore the relationship between control-signal and cutoff frequency really has to be exponential.

# Modular Components of the TimewARP 2600 4

## 4.1 Top Row Control Panel Buttons and Indicators



Just above the panel graphics, outside the “case” of the TimewARP 2600, is a horizontal row of buttons and indicators for patch storage, import/export, voice-cloning, and other operations.

These powerful features of the TimewARP 2600 have no equivalent in the world of analog synthesis; they are unique digital extensions of the original ARP 2600 synthesizer.

### 4.1.1 Patch Lock Button (Padlock Icon)

This padlock button helps you to avoid accidentally overwriting your favorite patch. Basically it disables the *Save* button, while leaving the *Save As* button enabled. Under these conditions, you can save your current patch only by assigning it a new name.

### 4.1.2 Group, Category, and Patch Drop-Down Lists

The TimewARP 2600 gives you a three-level hierarchy for storing and organizing your patches. All *Patches* are sorted into various *Categories*, which are in turn sorted into major *Groups*. Each of the three patch selection buttons generates a drop-down list associated with one layer in this hierarchy.

*Groups*, *Categories*, and *Patches* can also be selected by keyboard shortcuts. The up/down arrow keys on the computer keyboard select *Patches*, the left/right arrow keys move between *Categories*, and using the control key with the left/right arrow keys moves between *Groups*.

### 4.1.3 Save Button.

The *Save* button saves the current patch configuration and settings under the name of the most recently loaded patch. This button is disabled if the patch is locked (see 4.1.1 above).

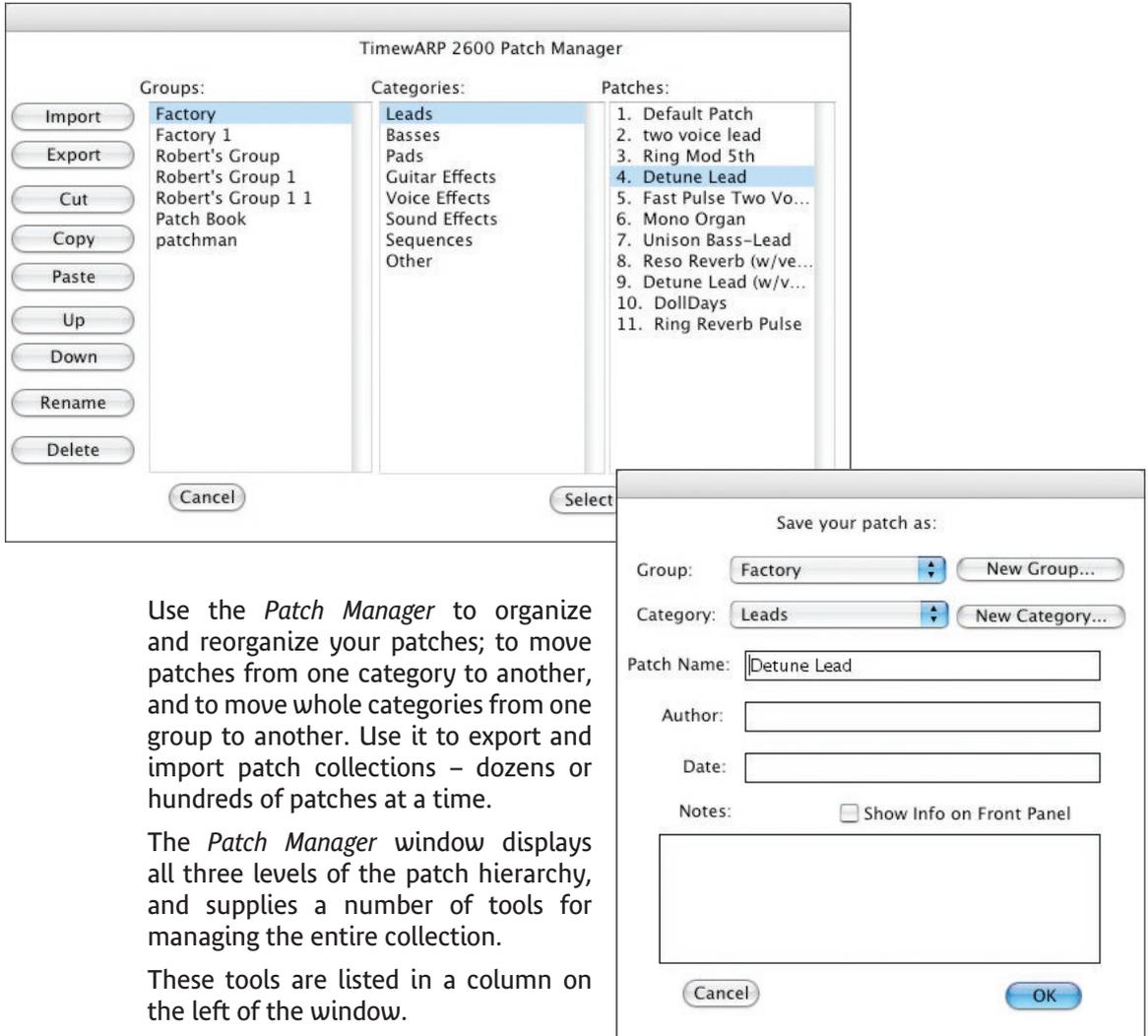
#### 4.1.4 Save As Button.

The *Save As* button saves the current patch configuration and settings under a group, category, and patch name of your choice.

Within the *Save As* dialog, you may create new groups and categories at will.

There is no limit to the number of groups and categories you may create.

#### 4.1.5 Patch Manager Button



Use the *Patch Manager* to organize and reorganize your patches; to move patches from one category to another, and to move whole categories from one group to another. Use it to export and import patch collections – dozens or hundreds of patches at a time.

The *Patch Manager* window displays all three levels of the patch hierarchy, and supplies a number of tools for managing the entire collection.

These tools are listed in a column on the left of the window.

To use them, *select* one or more items from the hierarchy, and then *click* on the operation you want to perform. Any operation that cannot be applied to the current selection of items will be disabled in the list.

#### 4.1.5.1 **Import / Export**

Use the *Import* / *Export* commands to write or read entire *Groups*, *Categories*, and *Patches* to/from external files.

*Export* works on whatever items are currently selected; by selecting all of the current groups, you can use *Export* to make a backup of all of your patches.

When you export a single *Patch*, the names of the *Group* and *Category* for the patch are exported with it.

*Import* asks you to select the file you want to read in. *Import* will never overwrite any existing *Group*, *Category*, or *Patch*; if any *Group* or *Category* or *Patch* in the file to be imported has the same name as a *Group* or *Category* or *Patch* that already exists, *Import* will append a number to the loaded name. So when a friend sends you his collection of a thousand patches, you can import them without worrying about possible name overlaps.

#### 4.1.5.2 **Cut / Copy / Paste**

The *Cut* / *Copy* / *Paste* buttons appear within the *Patch Manager* dialog to move items from one place in the hierarchy to another. You can, for example, *Cut* a patch, or a group of patches, from one category and *Paste* it into another.

#### 4.1.5.3 **Up / Down**

The *Up* / *Down* buttons appear within the *Patch Manager* dialog to move selected items *Up* or *Down* in their list. (This is useful for arranging your MIDI patch lists.)

#### 4.1.5.4 **Rename / Delete**

The *Rename* / *Delete* buttons appear within the *Patch Manager* dialog to *Rename* or *Delete* selected items.

#### 4.1.6 **Voice Button**

Clicking on the *Voice* button activates a drop-down list from which you may select the number of simultaneously sounding voices you want to use.

Because the TimewARP 2600 is a true analog synthesizer emulator, its modules are running even when no audio signals appear at the synthesizer's output. Each voice added to the multi-voice capability of the TimewARP 2600 is a clone of the entire patch and module set. This will have an immediate and obvious effect on the CPU load meter.

So: how many voices you can generate without overtaxing your CPU will depend on your machine's clock speed.

### 4.1.7 Reset Button

The *Reset* button removes all patch cords and returns all sliders to a standard position.

### 4.1.8 MIDI Indicator

This virtual LED glows when there is any MIDI input to the TimewARP 2600 – not just keystrokes, but also controller input and sysex dumps.

### 4.1.9 Output-Level Meter

This shows the output signal level. If it reaches into the red segment, your signal will distort.

### 4.1.10 CPU Load Meter

This meter shows, roughly, how much of the time between samples (the sample period) is being devoted to the TimewARP 2600 emulation process. In a complex patch, or a many-voice polyphonic performance, the meter may indicate overload; when this happens, it is likely that the TimewARP 2600 output signal will be interrupted, so your audio feed will develop a glitch. To avoid this, you will have to simplify your patch, or decrease the number of voices, or acquire a faster, more capable computer.

### 4.1.11 The Magic Logo



At the lower right of the main panel is the TimewARP 2600 Logo. Clicking on this brings up a menu:

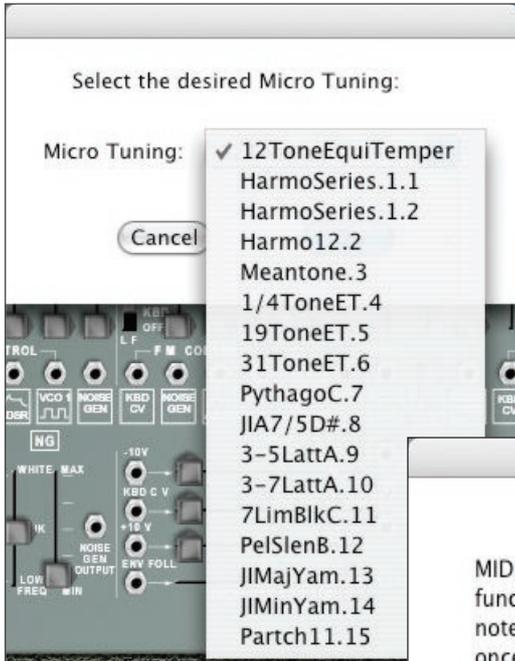
**4.1.11.1 About TimewARP 2600** identifies the team; the people who worked together to bring you this software.

#### 4.1.11.2 Load/Save MIDI Maps



Use the *Load/Save MIDI Maps* commands to save – and reload – the MIDI-controller to slider assignments that you set up. In the TimewARP 2600, these are *global* assignments, independent of any particular patch settings; saving a patch does not save these assignments, and loading a patch does not change the current assignments. You can, if you want, set your mappings once, and they will be there throughout all of your personal patch changes.

### 4.1.11.3 Load Microtuning



You may also load alternate tunings for the keyboard. These are described in Appendix 6.1. The TimewARP 2600 does not allow you to modify these tunings or to save or create new ones.

*Microtunings* are a global attribute of the keyboard; once loaded, the tuning will govern anything you play until you load a different one, regardless of your patch changes

### 4.1.11.4

#### MIDI Beat Synchronization

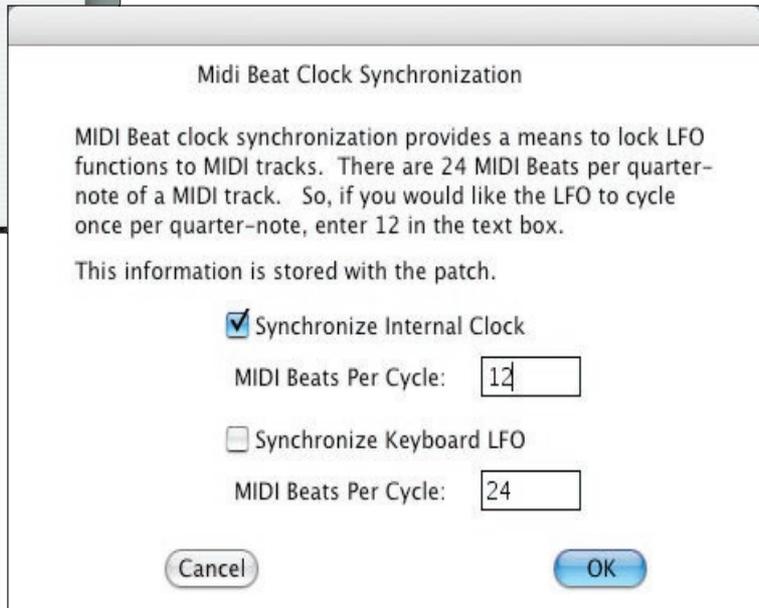
You may synchronize the *Internal Clock (IC)* (see section 4.13), to the *MIDI Beat Clock (MBC)* by specifying the number of *MBC* pulses per *IC* transition. As a reference, there are 24 *MBC* pulses per quarter note.

The keyboard *LFO* (see section 4.14.1) may also be synchronized to incoming the *MBC*, independently of the *Internal Clock*.

Setting different sync counts for these is a fun way to program complex rhythms that are locked to the tempo of your MIDI tracks.

In order for the *MBC* messages to be sent to the TimewARP 2600, you must enable *MIDI Beat Clock* in the Pro Tools MIDI Menu, and select the TimewARP 2600 as a recipient of these messages. Also, *MBC* messages are only sent when the Pro Tools transport control is running.

*MBC* synchronization is a *patch* attribute, *not* a *global* one; the sync counts you set here will be stored with the current patch when you save it.



### 4.1.12 Jacks, Patchcords, and Default Connections

The panel has eighty-one mini-jacks. Forty-five are inputs, twenty-nine are outputs, and 7 operate as both input and output.

Of the 45 inputs, 32 are in a row running across the center of the panel. (There are actually 34 jacks in the row, but the two labeled “gate” and “trig” are *outputs*.) This row of input jacks divides the control surface almost evenly in half.

Above this row, in the upper half of the control surface, there are only three *input* jacks. They are at the upper right corner, labeled *Left Input*, *Pan*, and *Right Input* respectively. All of the other jacks in the upper half of the control surface are *outputs*.

In the lower half of the control surface are inputs to the voltage processors, and of the column of four jacks in the section labeled *Sample & Hold*, the upper and lower jacks are inputs.

The seven jacks that are both input and output, belong to the *Electronic Switch* and the *Multiple* outlet. Because the switch works in either direction, it has either two inputs and one output or one input and two outputs. The *Multiple* output distributes at least one input to 1, 2, or 3 outputs.

All the remaining jacks are outputs. Most of them are labeled as such; a few are not, but have arrows pointing to them. For example, in the *Voltage Processor (VP)*, the three jacks furthest to the right are outputs; and in the *Envelope Generator* section of the upper half of the panel, the two jacks labeled *Gate* and *Trigger* are outputs.

### 4.1.13 Sliders and Slider Operations

There are fifty-eight sliders. Thirty-six of these are plain old signal attenuators. And of those 36, 29 are all in a row across the middle of the panel. (There are actually 31 sliders in the row, but the two on either side of the box labeled *Attack Release* are *not* plain old signal attenuators.)

Most of the attenuators are directly associated with either an input to something or an output from something. Each vertical attenuator across the middle of the panel, for example, adjusts the strength of the signal coming in from the input directly below it.

### 4.1.14 Normalled Jacks

The most commonly used signal connections are “normalled” (i.e. defaulted). A default signal is identified by a small icon at an input jack.

To see a simple example of this, note that the first input jack on the left, in the row of jacks running across the middle of the TimewARP 2600, is an input to the *Envelope Follower*. The symbol underneath this jack indicates that the default signal to this input comes from the *Preamp*. That means that the *Preamp* output is prewired to the *Envelope Follower* input, except when a plug is inserted into the jack.

For another simple example, note that in the same row of jacks, the third one from the right is a mixer input, and that the symbol just beneath it indicates that the default signal to this input comes from the *Voltage-Controlled Filter (VCF)*:

Note again that the fifth of the five audio inputs to the *VCF* is similarly defaulted from the *Noise Generator* (counting across from left to right this is the 21st jack):

So you can listen to this input by opening the *VCF* input to the mixer, and the *Noise Generator* input to the *VCF*. Now experimenting with the two horizontal control sliders at the top of the *VCF* panel will give you a wide range of filtered sounds.

It will be worth your while to experiment thoroughly and systematically with the default signal connections at this point, particularly if you are planning to use the TimewARP 2600 in live performance. In section 4 we will document the behavior of each separate module, and in section 5 we give sample patches for further experimentation; here we will only mention a few general principles to keep you from going out of your skull with complications:

Experiment with one signal at a time. With the *VCF*, for example, when you have listened to everything the filter can do with a noise input, close that input and open the default *VCO-3* sawtooth immediately to its left. Now you can experiment not only with the *VCF* controls, but also with the manual frequency controls of *VCO-3*; and when you have done that, experiment one by one with the control input signals to *VCO-3*.

## 4.2 Preamp/Gain Control



The *Preamp* section controls the gain of the audio signal(s) from the track in which the TimewARP 2600 is running. A rotary knob labeled *Gain* adjusts the signal level.

If the TimewARP 2600 is running in full stereo configuration – as a plug-in to a stereo track – the preamp will display two output jacks, one for each stereo channel. Use these signals for any purpose for which you might ordinarily use an internally-generated signal. You can filter them, run them through the *Ring Modulator*, or use one as an AM or FM program signal. The default input to the *Envelope Follower*, under these conditions, is taken from the left channel.

## 4.3 Voltage-Controlled Oscillators (VCO)

These generate three or more of the following basic waveforms. The output amplitude and phase relationships are the same for all oscillators. The oscillator sensitivity under virtual voltage control is 1vV/octave.

For convenience in fine-tuning control depth, the three attenuator-governed *FM Control* inputs at each oscillator provide three different sensitivity ranges. The leftmost slider is full-range; wide open, it passes its signal unchanged. The second slider is 50%; wide open, it passes its signal at half strength. The third slider, wide open, passes its signal at 25% of its original amplitude.

### 4.3.1 VCO 1



VCO 1 generates saw, square, and sine outputs. The sine output is a TimewARP 2600 extension; the original ARP 2600 VCO1 provided just sawtooth and squarewave outputs.

The default signal to the first (unattenuated) *FM Control* input is from the keyboard. The *Audio/LF* switch above this input switches the mode of the VCO from *Audio* (10Hz - 20,000Hz) to *LFO Mode* (0.03Hz – 30Hz). When the VCO is in *LFO Mode*, the default connection to the keyboard is removed. This can be overridden in this mode by patching a cable to the *Keyboard CV* output on the left side of the front panel.

The default signals to the next three *FM Control* inputs are from a) the *Sample & Hold*, b) the *ADSR Envelope Generator*, and c) *VCO2* sine.

### 4.3.2 VCO 2



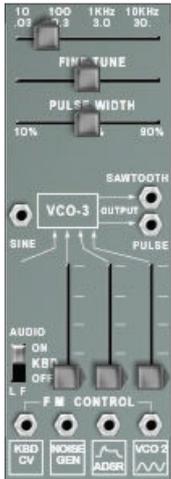
VCO 2 generates sine, triangle, sawtooth, and pulse outputs. A pulse-width slide control can adjust the duty cycle from 10% to 90%; at the middle of its travel, the pulse width is 50%, that is, a square wave.

The default signal to the first (unattenuated) *FM Control* input is from the keyboard. The *Audio/LF* switch above this input switches the mode of the VCO from *Audio* (10Hz - 20,000Hz) to *LFO Mode* (0.03Hz – 30Hz). When the VCO is in *LFO Mode*, the default connection to the keyboard is removed. This can be overridden in this mode by patching a cable to the *Keyboard CV* output on the left side of the front panel.

The default signals to the next three *FM Control* inputs are from a) the *Sample & Hold*, b) the *ADSR Envelope Generator*, and c) *VCO1* square.

There is a fourth attenuator-governed input, for digital control of the pulse width. The default signal at this PWM input is from the *Noise Generator*.

### 4.3.3 VCO 3

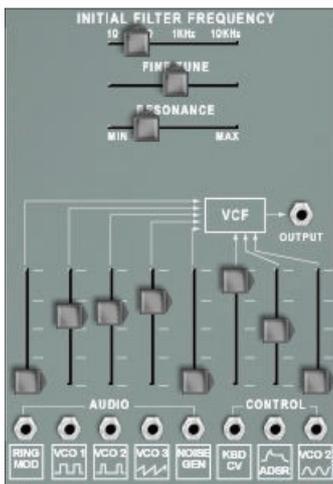


VCO 3 generates sawtooth, pulse, and sine outputs; the pulse width is manually variable. The sine output is a TimewARP 2600 extension; the original ARP 2600 VCO3 provided just sawtooth and pulse outputs.

The default signal to the first (unattenuated) *FM Control* input is from the keyboard. The *Audio/LF* switch above this input switches the mode of the VCO from *Audio* (10Hz - 20,000Hz) to *LFO Mode* (0.03Hz – 30Hz). When the VCO is in *LFO Mode*, the default connection to the keyboard is removed. This can be overridden in this mode by patching a cable to the *Keyboard CV* output on the left side of the front panel.

The default signals to the next three *FM Control* inputs are from a) the *Noise Generator*, b) the *ADSR Envelope Generator*, and c) *VCO2* sine.

### 4.4 Voltage Controlled Filter (VCF)



The *Voltage Controlled Filter* has variable cutoff frequency ( $F_c$ ) and resonance ( $Q$ ). The response below  $F_c$  is flat down to DC; above  $F_c$  the response falls off at 24Db per octave.  $F_c$  range is from 10Hz to 10KHz without control voltages; under voltage control,  $F_c$  can be driven as far down as 1 Hz and as high as 20KHz.

$F_c$  is controlled manually by a coarse tuning slider (labeled initial filter frequency) and a fine tune slider.  $F_c$  may also be controlled by external voltages; the sensitivity under voltage control is 1.0vV/oct.

The  $Q$ , or resonance, of the filter circuit is controlled by a single manual slider. As the  $Q$  is increased by moving this slider from left to right, the response below  $F_c$  is gradually attenuated until a sharp peak remains at the cutoff frequency. (Gain at  $F_c$  is always unity.)

At this  $Q$  setting, just below the point at which oscillation begins, the filter will ring distinctly in response to any sharply defined pulse presented to its signal input. In this state it is effectively analogous to

a highly resonant physical system, and may be used for various percussion effects depending on its resonant frequency (identical to  $F_c$ ) and on the impulse spectrum exciting it.

As the  $Q$  is raised still higher, beyond about the halfway point in the slider travel, the filter will oscillate. Operating in this state, it generates a pure sine wave. even in the absence of any signal input.

The VCF has five *Audio* signal inputs. They are fed through logarithmic attenuators to a summing point, and then to the VCF itself. The default input signals are from the *Ring Modulator*, *VCO-1 Square*, *VCO-2 Pulse*, *VCO-3 Sawtooth*, and *Noise Generator*.

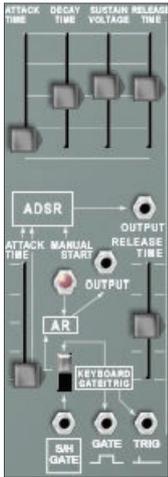
The VCF has three frequency *Control* inputs. The first is normally from the *Keyboard* pitch-control. The slider that governs this input is a TimewARP 2600 extension; on the original

ARP 2600, the keyboard control depth was not adjustable.

The second and third *FM Control* inputs are governed by linear attenuators; prewired to these are the *ADSR Envelope Generator* output and the *VCO-2 Sine* output.

Inserting a patch cord at an input jack automatically disconnects the default signal.

### 4.5 Envelope Generators



The *Envelope Generators* generate transient, positive-going waveforms, with controllable rise and fall times. They are used primarily with the *VCF* and *VCA*, in generating events whose time-varying spectrum and amplitude must be accurately and repeatably controlled. The output from each generator is a positive-going signal whose rise and fall time is set by slide controls on the generator, and whose onset and overall duration is determined by a gate signal.

The maximum value that either envelope can reach is +10vV; thus, unattenuated, either envelope is capable of driving a *VCF* or *VCA* from its minimum initial setting (10Hz for the *VCF*, -100Db for the *VCA*) all the way up to maximum. See 4.1.2 and 4.1.3, specifically the data on control input sensitivity. Reread too sections 2.1.6 through 2.1.7.

Gate signals for operating an *Envelope Generator* may originate with a *Manual Start* button, the *Keyboard Controller*, or any +10vV square-wave or pulse signal. The two-position switch just under the lower *AR* generator selects between the two latter sources. *The Manual Start* button overrides both of these.

#### 4.5.1 ADSR Envelope Generator

The *ADSR Envelope Generator* offers variable *Attack* time, initial *Decay* time, *Sustain* level, and final *Release* time. Four vertical sliders control these parameters: note that three of these are time parameters and the fourth – *Sustain* level – is not.

The generator produces an output only when a gate signal is present at its input. At the onset of a gate signal, the output level rises to +10vV, at a rate set by the *Attack* slider. When the level reaches +10vV, it immediately begins to fall, at a rate set by the *Decay* slider, to a level set by the *Sustain* slider. It remains at this level while the gate signal continues. Finally, when the gate ends, the output falls to zero at a rate set by the final *Release* slider.

If the gate ends at any time during the *ADSR* cycle, the generator immediately advances to its final *Release* phase: the current output level starts falling to zero from whatever value it had at the moment the gate ended.

When the *Envelope Generator* is controlled from the *Keyboard Controller*, it responds to the trigger signal from the keyboard – supposing that an envelope is already in process – by repeating the first two stages of an envelope and returning again to the *Sustain* level. For all other control gates presented to the generator input, circuitry internal to the generator itself derives a trigger signal from the leading edge of the gate.

### 4.5.2 AR Envelope Generator

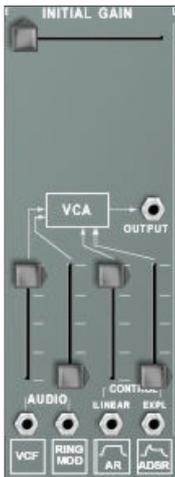
The *AR Envelope Generator* offers variable *Attack* time and final *Release* time. Two vertical sliders control these parameters.

The generator produces an output only when a gate signal is present at its input. At the onset of a gate signal, the output level rises to +10vV, at a rate set by the *Attack* slider. When the level reaches +10vV, it remains at this level while the gate signal continues. Finally, when the gate ends, the output falls to zero at a rate set by the final *Release* slider.

If the gate ends at any time during the *AR* cycle, the generator immediately advances to its final *Release* phase: the current output level starts falling to zero from whatever value it had at the moment the gate ended.

The *AR Envelope Generator* does not require a trigger signal for any phase of its operation.

### 4.6 Voltage Controlled Amplifier (VCA)



The *Voltage Controlled Amplifier* has a maximum gain of unity and a dynamic range of 100Db. With the initial gain control at maximum, and with no control input, the VCA will pass with unchanged amplitude any signal presented to its signal input. On the other hand, with the initial gain control at minimum, no signal will pass through the amplifier at all unless some *positive* signal level (the VCA does not respond to negative control signals) is present at one or both of its control inputs.

The first control input has *linear sensitivity*; the gain of the amplifier in response to a signal at this input is  $S/10$ , i.e. dividing the signal level by 10 will give the gain factor.

The second control input has *exponential sensitivity*; the gain of the amplifier in response to a signal at this input will equal 10Db/vV.

There are two audio signal inputs to the VCA. The default connections to these are from the *VCF* and the *Ring Modulator*. Inserting a patch cord automatically disconnects the default signal.

The default signal at the linear control input is from the *AR Envelope Generator*, and at the exponential input is from the *ADSR Envelope Generator*. Inserting a patch cord automatically disconnects the default signal.

## 4.7 Mix/Pan/Reverb Output Module



The three functions in this module provide final processing of the output signal. That, at least, is what they are intended for; you may actually use them in other roles, for any purpose you please.

If you leave the default connections undisturbed, the module is configured as a two-input Mixer, which feeds a *Pan Control* and a *Reverb* unit, which are themselves mixed to feed the final left/right system output channels.

When the TimewARP 2600 is configured for mono operation, this section omits the Pan control and provides only one output channel.

### 4.7.1 Mixer

The two inputs to this *Mixer* carry default connections from the *VCF* and the *VCA*; they can of course be overridden with patch cords.

The two jacks just above the sliders and below the *Mixer* graphic are *not* inputs; they are the *outputs* from the attenuators. This lets you use the two sliders as “floating” attenuators, in any situation where you need to set the strength of a signal. (Although if you do this, there’s no way to get any signals into the mixer.)

### 4.7.2 Pan Control

The *Pan Control* takes its input from the jack just below the horizontal pan slider. Normally, this signal comes from the mixer. Centered, the *Pan* feeds its input signal equally to the left and right channel outputs; moving the slider left or right shifts the signal balance accordingly between the two output channels.

### 4.7.3 Reverb Unit

The input to this unit is the rightmost jack in the row that runs across the middle of the panel. By default, it carries the *Mixer* output. The output jack, to its upper right, provides a 100% wet signal from the *Reverb*, at a fixed level. (There are interesting patches in which this signal is subjected to further processing via, say, the *Ring Modulator* or the *Envelope Follower*.)

The two sliders adjust the wet-dry mix fed to each output channel.

## 4.8 Envelope Follower



The *Envelope Follower* generates, from any audio-frequency input, a fluctuating DC output level directly proportional to the average moment-by-moment input signal amplitude. Its sensitivity is such that, with the input attenuator wide open, a 1V P-P square wave will produce a +10vV output. The maximum output is +10vV.

The risetime, or time it takes for the *Envelope Follower* to respond to any sudden change in the amplitude of the signal input, is 10 milliseconds to 50% of final value and 30 milliseconds to 90% of final value.

Like all similar circuitry, the *Envelope Follower* tends to “ride” on low audio frequencies as if they themselves represented changes in signal amplitude; this is not critical, but has been held to a ripple of less than 1% P-P down to 100Hz and less than 10% down to 40Hz.

The primary use of the *Envelope Follower* is with external instruments. Essentially it extracts, from any audio input, a control signal representing the amplitude-envelope of that input: this signal may control the *VCF*, *VCA*, or any of the *VCO*'s. The *Envelope Follower* output is an envelope and can be used in the same fashion as the output from either of the envelope generators.

The default input to the *Envelope Follower* is from the preamplifier. When the TimewARP 2600 is configured for stereo input, it is the first (left) channel preamplifier output.

## 4.9 Ring Modulator



The *Ring Modulator* is essentially a multiplier; from its two inputs A and B it produces the output function  $A \times B / 5$ . The kind of transformation this effects on input signals depends to a large extent on what they are and on whether the modulator is AC or DC coupled to them. This is selected by the *Audio/DC* switch at the bottom of the modulator.

When the inputs are AC coupled (*Audio* position of the switch), any DC component present in them is canceled before they are fed to the modulator. Thus a sawtooth that starts from zero and goes to +10vV will instead start at -5vV and move to +5vV so that its overall positive and negative deviation cancels to zero. Under these conditions the modulator will generate from any two periodic signals an output signal consisting of the sum and difference frequencies that can be generated from the frequencies of the two inputs. The input frequencies themselves will be suppressed.

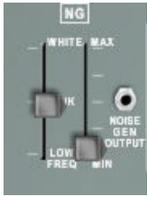
If both signals are audio-frequency, a large variety of harmonic and inharmonic timbres can be produced from the modulator, depending on the ratio of the input frequencies and on their own harmonic content. If A is a sine wave and we represent its frequency by  $F_a$ , and B is a complex waveform of frequency  $F_b$  with overtones  $2F_b$ ,  $3F_b$ ,  $4F_b$ , etc., then the output of the modulator will be a complex waveform with frequency components  $F_b + F_a$ ,  $F_b - F_a$ ,  $2F_b \times F_a$ ,  $3F_b \times F_a$ ,  $4F_b \times F_a$ , etc. A moment's experimentation with the prewired sawtooth and sine inputs to the modulator will demonstrate the complexity of the timbres that can be generated by this simple means.

If, still with AC coupling, one input is subsonic and the other at some audio frequency, there will be an output from the modulator only when the value of the subsonic input is changing, and the output will be roughly proportional to the rate of change. If, for example, the subsonic input is a square wave, the modulator output will be a series of short, decaying tonebursts – one at each rise or fall in the input signal.

When the inputs are DC coupled, any DC component in either one of the inputs will pass into the modulator and affect the modulating process. The effect when both inputs are at audio frequency is to allow into the output waveform some of the input frequencies in addition to the sum and difference frequencies. The effect when one of the inputs is subsonic is that the modulator operates as a voltage-controlled amplifier: the output amplitude will be in direct proportion to the instantaneous amplitude of the low-frequency input and will vary as its absolute value varies. Also, the output phase will reverse when the low-frequency input signal changes from positive to negative or vice versa.

The AC-coupling time constants are 235 msec, for the left input and 90 msec, for the right input.

#### 4.10 Noise Generator (NG)

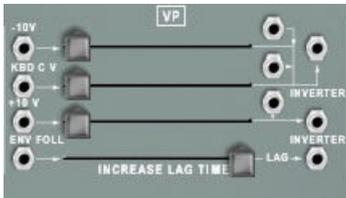


The *Noise Generator* has two manual controls: one for spectral balance (“color”) and one for output level.

The spectral balance is continuously variable from white to red (low-frequency noise output). In the latter case the output falls off at the rate of 6Db/Octave; the pink noise position approximates a -3Db/Octave slope.

The level control, at minimum, cuts off the output signal completely. At maximum, the output is clipped at 20vV P-P to produce binary, or two-valued, noise. Clipping begins with the level control approximately half open. input.

#### 4.11 Voltage Processors (VP)



The *Voltage Processors* are simple utility functions for mixing, inverting, and shaping signals.

##### 4.11.1 VP #1

*VP #1* has four signal inputs and one output. Two of the inputs have attenuators. The output signal is the inverted sum of all four inputs.

The attenuator-governed inputs carry default connections from +10vV and from the keyboard pitch-control. Opening the +10vV slider thus produces up to -10vV at the processor output.

##### 4.11.2 VP #2

*VP #2* has two signal inputs and one output. One of the inputs has an attenuator. The output signal is the inverted sum of the two inputs.

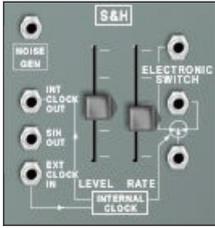
The attenuator-governed input carries a default connection from -10vV, so that opening the slider produces up to +10vV at the output.

##### 4.11.3 The Lag Processor

The *Lag Processor* is a low-pass filter for processing control signals. The slider adjusts its cutoff frequency. The corresponding rise-time ranges from 0.5ms with the slider at minimum, to 500msec – about half a second – with the slider at maximum.

The *Lag Processor* can be used to process audio signals, as a -6Db/octave manual filter with a maximum Fc of approximately 1KHz.

### 4.12 The Sample & Hold Module (S/H)



The *Sample & Hold Module* produces stepped output signal levels, by sampling the instantaneous value of any signal at its input. The stepped levels produced in this manner are useful for controlling oscillator and filter frequencies and – occasionally – *VCA* gain.

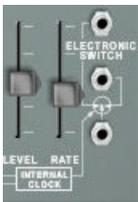
The *S/H* circuit has a *signal* input (the waveform to be sampled), a *trigger* input, and an *output* giving the result of the sampling operation. The trigger input is defaulted from the internal clock, but any square or pulse wave, or the keyboard gate or trigger signals, will work.

Upon being triggered, the *S/H* sets its output level to the same value as the input signal at that instant. After the trigger, the output signal will hold that level until the next trigger pulse.

Any signal whatsoever may be sampled. The default input is from the *Noise Generator*, so that the step sequence is random. The accompanying diagrams show how, when the signal being sampled is random noise, the output voltages are correspondingly unpredictable. An infinite variety of cyclical output patterns may be obtained, on the other hand, by sampling a periodic waveform. Different ratios of the sampling frequencies to the frequency of the waveform being sampled create different melodic patterns (if the output level is controlling a *VCO*).

The level control attenuates the input signal before it is fed to the *S/H* circuit. The rate control actually belongs to the internal clock; when that is disconnected from the *S/H* circuit, the rate control has no effect on the operation of the *S/H* circuit.

### 4.13 The Internal Clock / Electronic Switch



The *Internal Clock* is a manually controlled low-frequency square-wave oscillator. It is the default trigger source for the *S/H* device. It is also hardwired as the clock source for the *Electronic Switch*.

Under MIDI control, the *Internal Clock* may be synchronized to incoming *MIDI Beat Clocks*; see section 4.1.11.4.

The *Electronic Switch* has two connections on one side and one on the other, as indicated by the panel graphics. For clarity, let's call these three jacks A-1, A-2, and B. The switch alternates between connecting A-1 to B, and A-2 to B. It doesn't matter which side is the signal source and which is the destination; the switch works the same regardless.

The switching rate is governed by the *Internal Clock*. This is a permanent feature of the switch.

## 4.14 The Virtual Keyboard



This has five octaves, 60 keys. The control panel, at the left, is modeled on but not identical to the original ARP 3620 keyboard.

### 4.14.1 Low Frequency Oscillator (LFO) Section

The keyboard unit has its own *LFO* section, independent of any of the standard *VCO*'s. It can be used in two ways: for vibrato, or for automatically repeated keyboard gates (as, for example, in imitating the repeated notes of a mandolin). Three sliders govern the *Speed*, *Delay*, and *Depth* of the *LFO*.

Under MIDI control, the keyboard *LFO* may be synchronized to incoming *MIDI Beat Clock* signals; see section 4.1.11.4.

### 4.14.2 Dual-Pitch Control Output

Like the original ARP 2600, the TimewARP 2600 virtual keyboard can generate a second pitch-control signal when two keys are depressed. This signal is available at the two jacks labeled *Upper Voice* at the lower left of the keyboard module.

To use one of these, simply patch it to an oscillator. That oscillator will now track the uppermost key depressed rather than – as with the standard keyboard control signal – the lower key.

### 4.14.3 Gate and Trigger Control

Two switches in the upper right quadrant of the keyboard module govern the logic of the keyboard gate and trigger signals.

When the *Trigger Mode* switch is set to *Off*, the keyboard generates a continuous gate signal as long as any key is depressed, and generates a trigger signal only on the transition from no key depressed to any key depressed. In this operating mode, you have complete performing control over the production of trigger signals; to avoid them, play legato, and to generate them, play non-legato. This is the baseline logic of the original ARP 2600 keyboard.

With this switch set to *On*, the keyboard will generate a trigger on every new keypress, regardless of your performing habits. The gate logic is not affected.

The three-position switch labeled *Auto Repeat* is *Off* in its center position. This is the default.

In its lower position, the keyboard gate and trigger are taken from the local *LFO*. Actual key depressions no longer play a role in gating. In its upper position, the *LFO* and the keypress are ANDed together; when you press a key, there is a series of pulses from the *LFO*, and when you release the key, the series stops. This is the mandolin effect we mentioned above.

The keyboard *Gate* and *Trigger* signals are available on the main panel, from two jacks in the *Envelope Generator* section.



# Patching the TimewARP 2600

# 5

Please see the *Patchman.pdf* file found on the install disc or at [www.wayoutware.com](http://www.wayoutware.com) for suggestions. If you are new to audio synthesis, consult the tutorial patch collection.



# Appendices 6

## **6.1 Table of Alternate keyboard tunings**

Tuning Presets, compiled by Robert Rich

### **6.1.1 12 Tone Equal Temperament (non-erasable)**

The default Western tuning, based on the twelfth root of two. Good fourths and fifths, horrible thirds and sixths.

### **6.1.2 Harmonic Series**

MIDI notes 36-95 reflect harmonics 2 through 60 based on the fundamental of A = 27.5 Hz. The low C on a standard 5 octave keyboard acts as the root note (55Hz), and the harmonics play upwards from there. The remaining keys above and below the 5 octave range are filled with the same intervals as Carlos' Harmonic 12 Tone that follows.

### **6.1.3 Carlos Harmonic Twelve Tone**

Wendy Carlos' twelve note scale based on octave-repeating harmonics.

A = 1/1 (440 Hz).

1/1 17/16 9/8 19/16 5/4 21/16 11/8 3/2 13/8 27/16 7/4 15/8

### **6.1.4 Meantone Temperament**

An early tempered tuning, with better thirds than 12ET. Sounds best in the key of C. Use this to add an authentic touch to performances of early Baroque music. C = 1/1 (260 Hz)

### **6.1.5 1/4 Tone Equal Temperament**

24 notes per octave, equally spaced  $24\sqrt[24]{2}$  intervals. Mexican composer Julian Carillo used this for custom-built pianos in the early 20th century.

### **6.1.6 19 Tone Equal Temperament**

19 notes per octave ( $19\sqrt[19]{2}$ ) offering better thirds than 12 ET, a better overall compromise if you can figure out the keyboard patterns.

**6.1.7 31 Tone Equal Temperament**

Many people consider 31root2 to offer the best compromise towards just intonation in an equal temperament, but it can get very tricky to keep track of the intervals.

**6.1.8 Pythagorean C**

One of the earliest tuning systems known from history, the Pythagorean scale is constructed from an upward series of pure fifths (3/2) transposed down into a single octave. The tuning works well for monophonic melodies against fifth drones, but has a very narrow palate of good chords to choose from. .

C = 1/1 (261.625 Hz)

1/1 256/243 9/8 32/27 81/64 4/3 729/512 3/2 128/81 27/16  
16/9 243/128

**6.1.9 Just Intonation in A with 7-limit Tritone at D#**

A rather vanilla 5-limit small interval JI, except for a single 7/5 tritone at D#, which offers some nice possibilities for rotating around bluesy sevenths.

A = 1/1 (440 Hz)

1/1 16/15 9/8 6/5 5/4 7/5 3/2 8/5 5/3 9/5 15/8

**6.1.10 3-5 Lattice in A**

A pure 3 and 5-limit tuning which resolves to very symmetrical derived relationships between notes. A = 1/1 (440 Hz)

1/1 16/15 10/9 6/5 5/4 4/3 64/45 3/2 8/5 5/3 16/9 15/8

**6.1.11 3-7 Lattice in A**

A pure 3 and 7-limit tuning which resolves to very symmetrical derived relationships between notes. Some of the intervals are very close together, offering several choices for the same nominal chords.

A = 1/1 (440 Hz)

1/1 9/8 8/7 7/6 9/7 21/16 4/3 3/2 32/21 12/7 7/4 63/32

**6.1.12 Other Music 7-Limit Black Keys in C**

Created by the group Other Music for their homemade gamelan, this offers a wide range of interesting chords and modes.

C = 1/1 (261.625 Hz)

1/1 15/14 9/8 7/6 5/4 4/3 7/5 3/2 14/9 5/3 7/4 15/8

**6.1.13 Dan Schmidt Pelog/Slendro**

Created for the Berkeley Gamelan group, this tuning fits an Indonesian-style heptatonic Pelog on the white keys and pentatonic Slendro on the black keys, with B and Bb acting as 1/1 for their respective modes. Note that some of the notes will have the same frequency. By tuning the 1/1 to 60 Hz, Dan found a creative way to incorporate the inevitable line hum into his scale.

Bb, B = 1/1 (60 Hz)

1/1 1/1 9/8 7/6 5/4 4/3 11/8 3/2 3/2 7/4 7/4 15/8

**6.1.14 Yamaha Just Major C**

When Yamaha decided to put preset microtunings into their FM synth product line, they selected this and the following tuning as representative just intonations. As such, they became the de-facto introduction to JI for many people. Just Major gives preferential treatment to major thirds on the sharps, and a good fourth relative to the second.

C = 1/1 (261.625)

1/1 16/15 9/8 6/5 5/4 4/3 45/32 3/2 8/5 5/3 16/9 15/8

**6.1.15 Yamaha Just Minor C**

Similar to Yamaha=92s preset Just Major, the Just Minor gives preferential treatment to minor thirds on the sharps, and has a good fifth relative to the second.

C = 1/1 (261.625)

1/1 25/24 10/9 6/5 5/4 4/3 45/32 3/2 8/5 5/3 16/9 15/8

**6.1.16 Harry Partch 11-limit 43 Note Just Intonation**

One of the pioneers of modern microtonal composition, Partch built a unique orchestra with this tuning during the first half of the 20<sup>th</sup> century, to perform his own compositions. The large number of intervals in this very dense scale offers a full vocabulary of expressive chords and complex key changes. The narrow spacing also allows fixed-pitched instruments like marimbas and organs to perform glissando-like passages.

G = 1/1 (392 Hz, MIDI note 67)

1/1 81/80 33/32 21/20 16/15 12/11 11/10 10/9 9/8 8/7 7/6

32/27 6/5 11/9 5/4 14/11 9/7 21/16 4/3 27/20 11/8 7/5

10/7 16/11 40/27 3/2 32/21 14/9 11/7 8/5 18/11 5/3 27/16 12/7

7/4 16/9 9/5 20/11 11/6 15/8 40/21 64/33 160/81



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